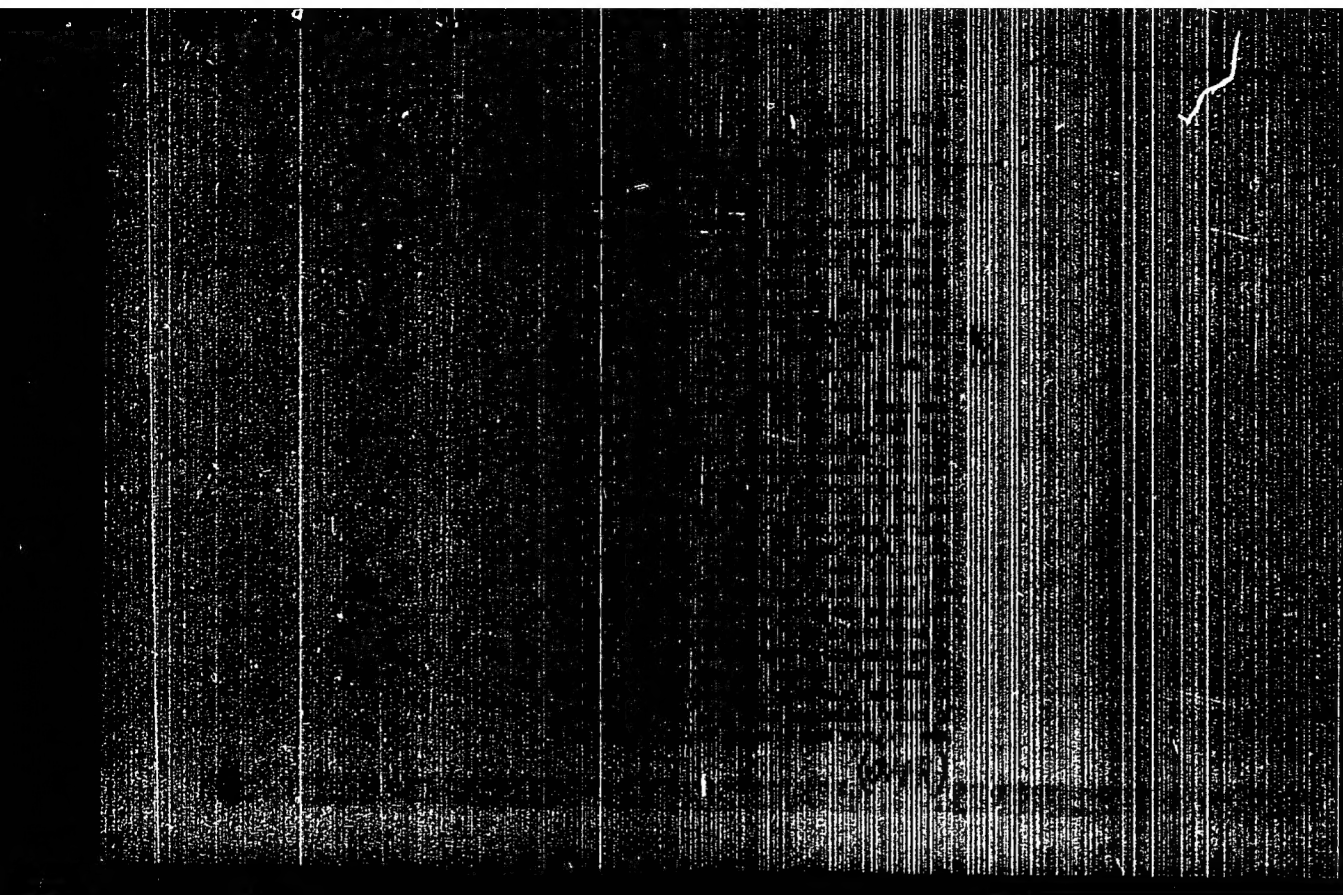


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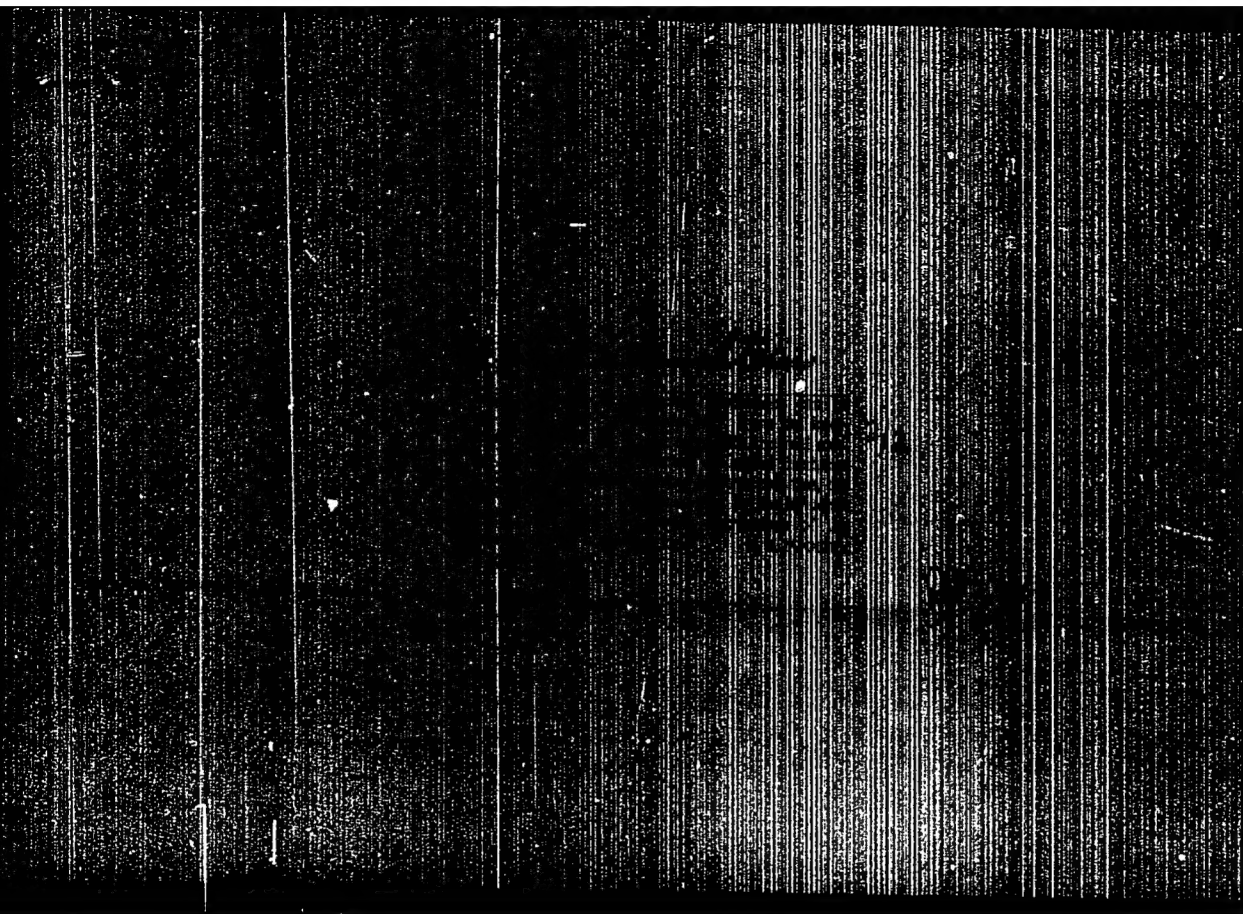


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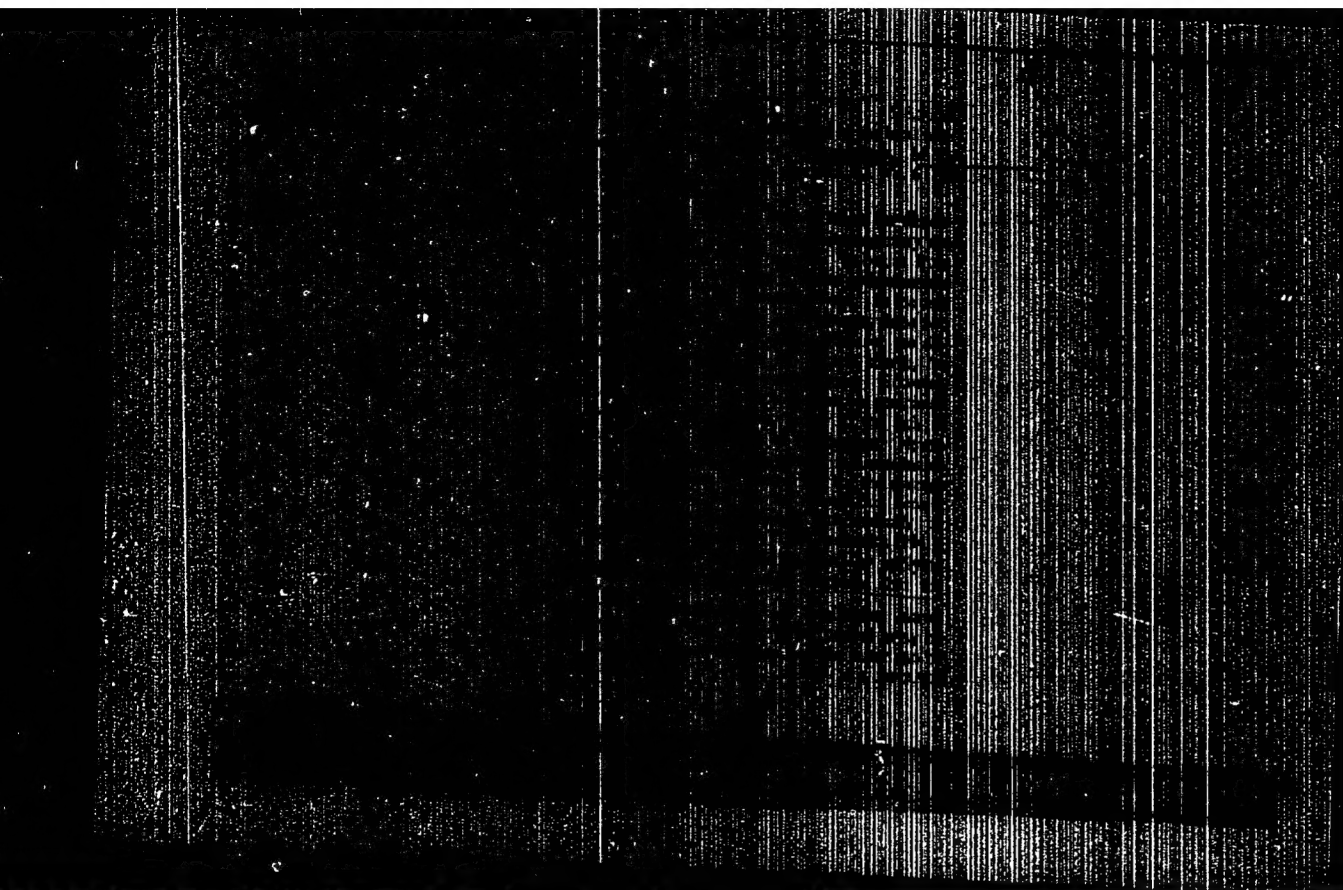


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*GERTSENSHTEYN, M.Ye.*

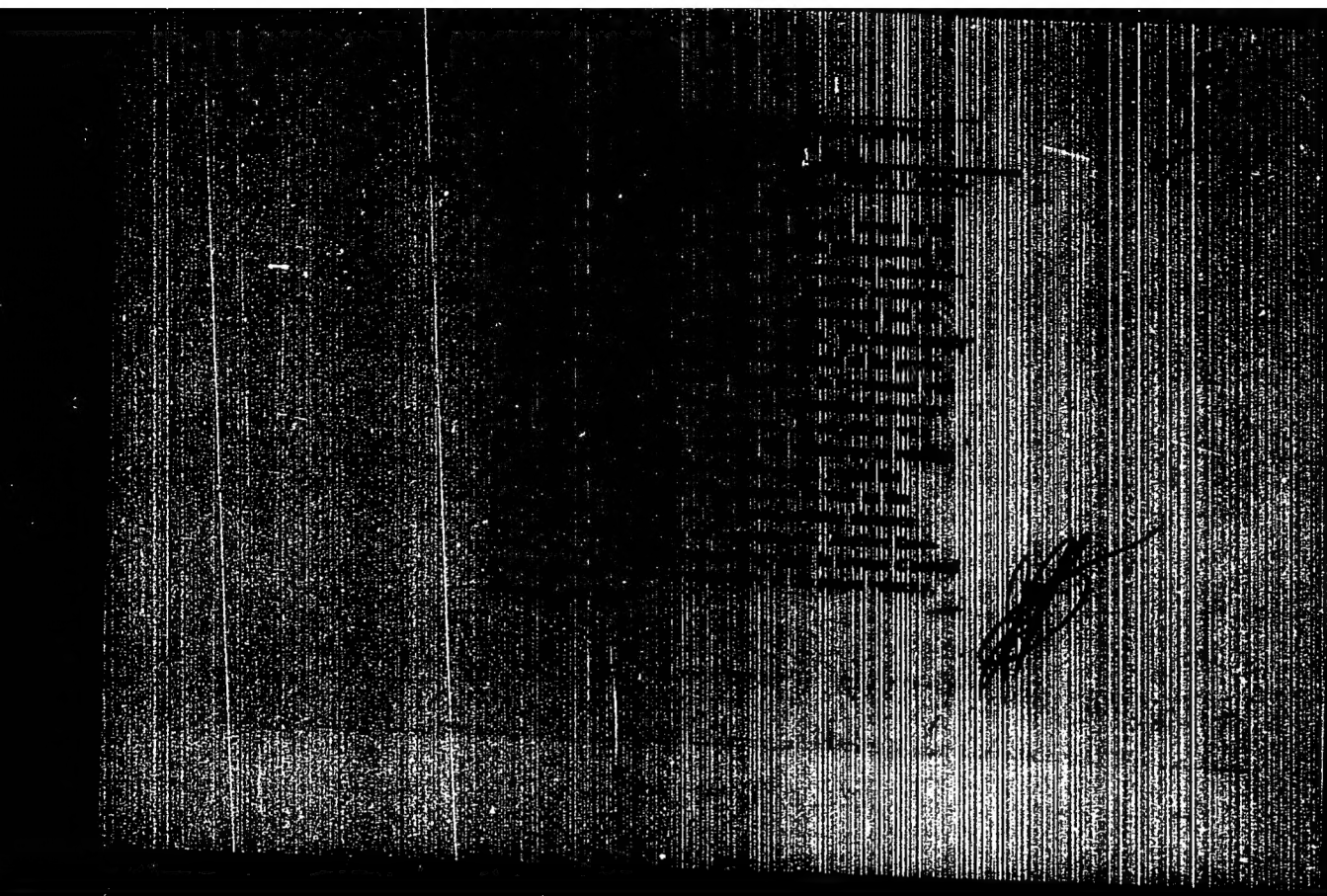
POTEMKIN, V.V.; GERTSENSHTEYN, M.Ye.

G.V.Gordeev's strata theory. Zhur.eksp. i teor.fiz. 24 no.5:610-612  
My '53. (MLBA 7:10)

(Nuclear physics)

"APPROVED FOR RELEASE: 09/24/2001

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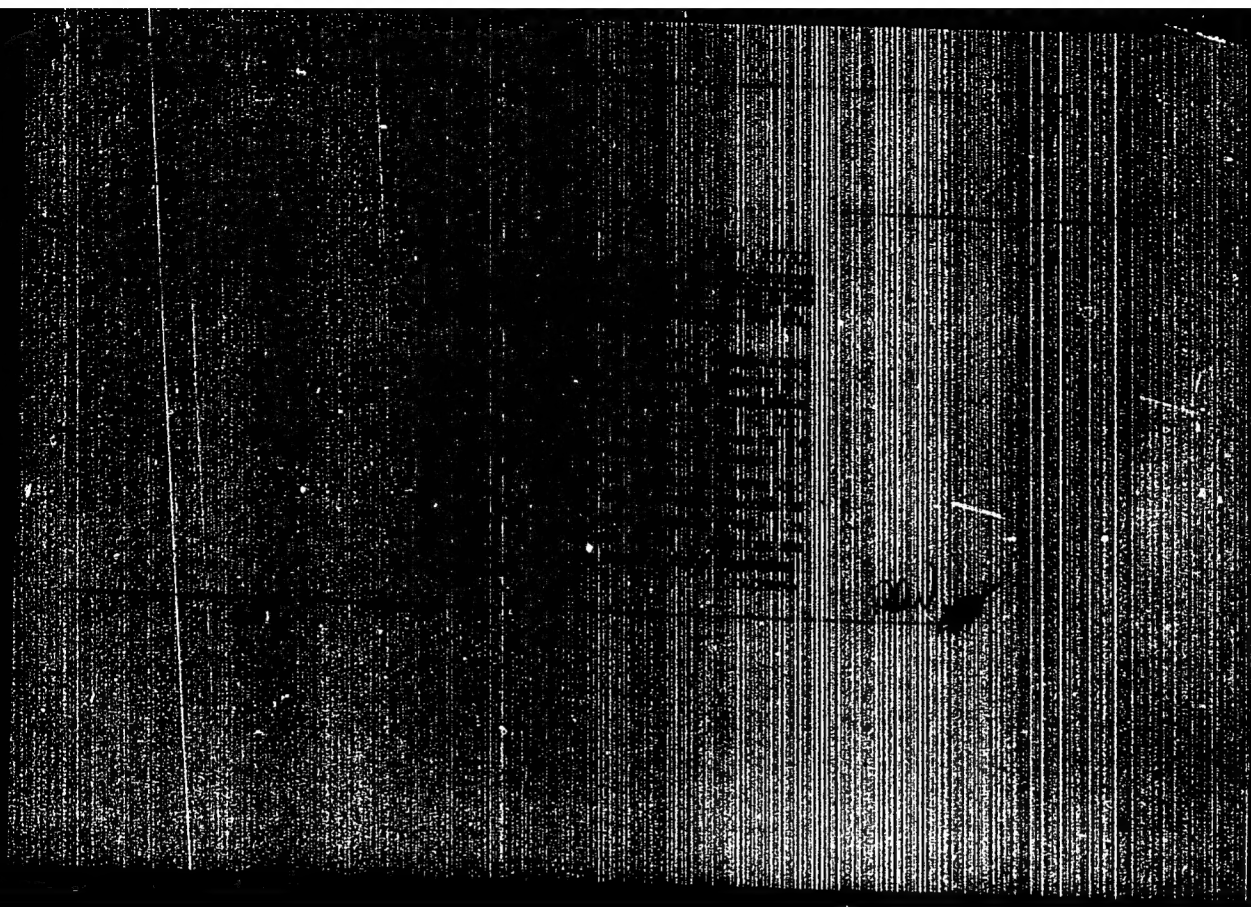


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APPROVED FOR RELEASE: 09/24/2001

CIA-RDP86-00513R000514920017-2"

GERTSENSHTEYN, M. Ye.

USSR/Physics - Self-excited oscillations

FD 405

Corn 1/1

Author : Gertsenshteyn, M. Ye.

Title : Self-excited oscillations in gaseous discharge at high pressures

Periodical : Usp. eksp. i teor. fiz. 36, 54-63, Jan 1954

Abstract : Treats the interaction of sound waves and electron waves in gas-discharge plasma. Demonstrates the possibility of self-excited oscillations for a definite interval of frequencies. Thanks V. V. Potemkin for his judgment of the physical results. Fourteen references, including K. F. Teodorovich, avtokolebani'nyye sistemy (Self-excited oscillator systems), State Technical-Theoretical Literature Press, 1952.

Institution : Moscow State University

Submitted : November 1, 1953

GERTSENHTEYN, M. Ye.  
USSR Physics - Electrodynamics

FD-115

Card 1/1 : Pub 146-3/1c

Author : Gertsenshtein, M. Ye.

Title : Energy current in spatial dispersing media

Periodical : Zhur. eksp. i teo. fiz., 26, 580-583, Jun. 1954.

Abstract : S. M. Rytov's results (ibid. 17, 930 (1947)) are generalized to the case of spatial dispersion when the partial derivative is not zero. It is shown that in this case the velocity of energy propagation coincides with the group velocity. 4 references. Indexed to V. V. Potemkin.

Institution : --

Submitted : October 15, 1952



GERTSENSHTEYN, M. Ye.  
USSR/Physics - Plasma

FD-795

Card 1/1      Pub. 146-8/21

Author : Gertsenshteyn, M. Ye.

Title : Dielectric permeability of plasma located in a stationary magnetic field

Periodical : Zhur. eksp. i teor. fiz., 27, 180-183, Aug 1954

Abstract : The tensor of the complex dielectric permeability of an electron gas is computed taking into account the thermal motion of electrons. Indebted to V. V. Potemkin. Sixteen references, including 3 foreign

Institution : Central Scientific Research Institute of Radio Measurements

Submitted : October 15, 1953

GERTSENSHTEIN, M. YE.

✓ Low-frequency oscillations in the positive column of a glow discharge. M. E. Gertsenshtein and V. V. Potemkin (Moscow State Univ.). *Zhur. Eksp. i Teor. Fiz.* 27, 642-64 (1954).—It is assumed that the phase delay is caused by longitudinal electromagnetic waves propagated along the axis of the pos. column. The internal resistance of a discharge tube as a wave generator is of the order of several hundred ohms. An analysis of luminescent phenomena in the discharge shows that there is a connection between the waves and the current pulses; the periodic luminous structure disappears on lowering the pressure at a pressure  $p_{osc}$ . An equiv. circuit is developed for the discharge tube acting as a pulse generator. The amplitude and the frequency of pulsation are changing periodically with the anode-cathode distance. 62

M. Potemkin

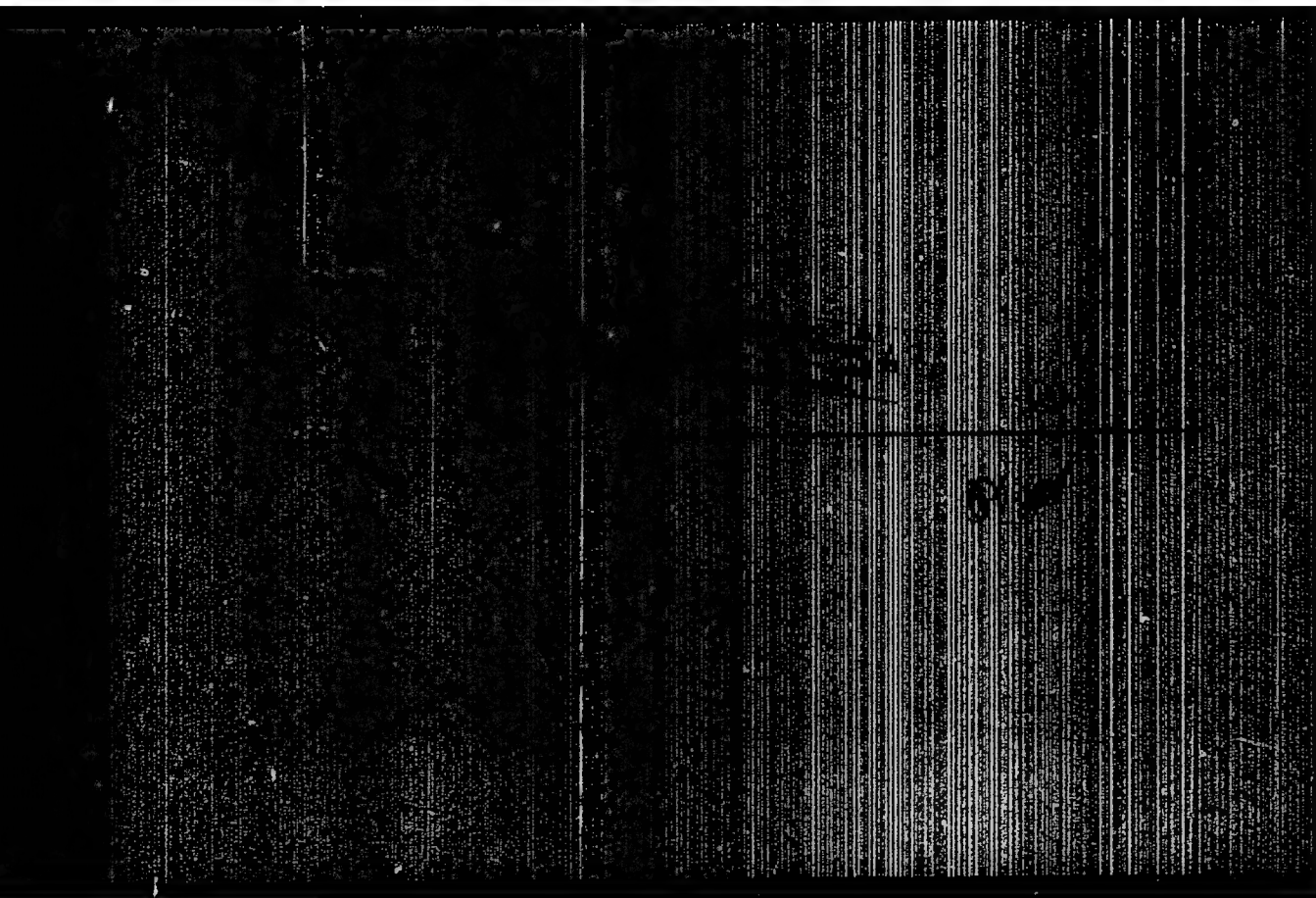
①

GERTSENHTEYN, M.Ye.

✓ Influence of elastic collisions of electrons and ions on longitudinal electric waves in plasma. M. B. Gertsenshtein (Moscow State Univ.). *Zhur. Fiz. i Teor. Fiz.* 6:27, 662-6 (1964).—A simplified form is developed for the collision integral. Two cases are discussed: (1) that when the phase velocity of the longitudinal wave is large compared to thermal velocity and (2) that when the phase velocity is small. In the 2nd case the losses are so small that a small quantity of energy will cause oscillation by autoexcitation. S. Pakunov

"APPROVED FOR RELEASE: 09/24/2001

CIA-RDP86-00513R000514920017-2



APPROVED FOR RELEASE: 09/24/2001

CIA-RDP86-00513R000514920017-2"

GERTSENSHTEYN, M.Ye.

Correlation of fluctuations in electron gases. Zhur. tekhn. fiz. 25  
no.5:834-840 My '55. (MLRA 8:7)  
(Electrons)

GERTSENSHTEYN, M.Ye.; BRYANSKIY, L.N.

Attenuator errors due to disagreement in the path of superhigh  
frequencies. Izv.tekh. no.1:28-33 Ja-F '56. (MLRA 9:5)  
(Radio, Shortwave) (Wave guides)

GERTSENSHTEYN, M.Ye.

Determining the shunting conductivity of the probe in recording  
circuits. Izv.tekh. no.4:37-38 J1-Ag '56. (MLRA 9:11)  
(Electric measurements)

GERTSENH'EYN, M.Ye.; JRYANSKIY, L.N.

Eliminating phase distortions in power measurements. Izv.tech.  
no.6:40-43 N-D '56. (MIRA 10:1)  
(Electric measurements)



AUTHOR : Gertsenshteyn, M.E. and POKRAS, A.M.

"Wave Guide Splitter with Variable Coupling."

A-U Sci Conf dedicated to "Radio Day," Moscow, 20-25 May 1957.

PERIODICAL: Radiotekhnika i Elektronika, Vol. 2, No. 6, pp. 1221-1224,  
1957, (USSR)

ALL INFORMATION CONTAINED HEREIN IS UNCLASSIFIED

DATE 10-10-2001 BY 60322 UCBAW/STP

AUTHOR: Lertsenskiy, V. Ye.

Doc ID: A66-100-17

TITLE: Precision Electronic Voltmeter for relative measurements  
'Totsionnyy lampovyy voltmeter dlya otnositel'nykh izme-  
reniy'

DATE: 1961: Izmeritel'naya tekhnika, 1961, no. 3, pp. 7-12, 10 figs.

ABSTRACT: Measurements of **audio frequency voltages** are in most cases relative: a signal of constant frequency and form is fed to the voltmeter and the relations of the amplitudes are measured. The linearity of the amplitude characteristic is, however, insufficient. The rectifying process of the a-c **voltage** is accompanied by non-linear distortions. A new voltmeter has been developed, therefore, the circuit diagram of which is shown in figure 1. The kенatron 6Д5 is used as a rectifying tube. The amplitude value of the sinusoidal **voltage** on the grid is 60 v. The measuring circuit for checking the linearity is given in figure 2. The results of the measurements de-

Cont 1/2

10/11/58-11/10/58

Precision Electronic Voltmeter for Relative Measurements

demonstrated that the error in the section of 50-100 divisions does not exceed 0.2%.

There are 2 graphs, 2 diagrams and 1 reference. 2 of which are Soviet and 1 German.

Card 2 2

ANT. CR: Vertol'matov, N.Ye. and Bryukhov, L.N. 103-3-147.

TITLE: A phase-shifter having a low Reflection Coefficient  
(Nizkoye otrazheniye volnovodnyy fazovyye shifrovannyye)

PERIODICAL: Radiotekhnika i Elektronika, 1968, Vol 11, No 5,  
pp 710 - 721 (USSR)

ABSTRACT: The standing-wave ratio of a terminating load in a waveguide can be measured either by means of a movable probe, or by a fixed probe and a phase-shifter. The first method is not suitable for the measurement of small standing-wave ratios (SWR) since its accuracy is comparatively low. A higher accuracy can be achieved by employing the phase-shifter method; the equipment necessary for these measurements consists of (see Fig. 1): 1) A ultra-high-frequency oscillator; 2) A matching transformer; 3) A fixed detector load; 4) A phase-shifter and, 5) the load. It is shown, however, that when there is small reflection, the phase-shifter is subject to the following errors: phase errors due to the losses in the phase-shifter, reflections from the movable elements of the shifter, errors due to the mis-matching of the oscillator and the driving section of the probe. The errors due to the reflections at the elements of the phase-shifter are analyzed in detail. It is concluded that the phase-shifter consists of

10-3-8-1-17

Waveguide phase-shifter having a low reflection coefficient

Dielectric plate whose thickness is  $s$  and height is  $h$ ; the permittivity of the material of the plate is  $\epsilon$  and the plate is located in the free space. The conditions are expressed by:

$$\frac{s}{h} \ll 1; \quad \frac{2\pi s}{\lambda} \ll 1; \quad \lambda = \frac{\lambda_0}{\sqrt{\epsilon}} \quad (8)$$

where  $\lambda_0$  is the wavelength in free space. If it is assumed that the material of the plate is isotropic, the boundary conditions at the plate can be written as Eq.(10) where  $E'$  and  $D'$  are the electric field and electric induction in the plate. The analysis of the conditions in the system can be carried out by solving Eq.(11), in which  $A$  defines a vector potential. Solution of Eq.(11) is in the form of a series expressed by:

$$A(x, y, z) = \sum_n a_n(z) A_n^s(x, y) \quad (14)$$

where the coefficients  $a_n$  can be obtained by solving a set of ordinary differential equations, as expressed by Eqs.(15), in which  $\epsilon_n$  is given by Eq.(10). Eq.(15) can be solved by the

method of successive approximations and in the first approximation

Waveguide Phase Shifter Having a Low Reflection Coefficient

they can be expressed in the form of Eqs.(19). Substituting Eq.(19) in the form of Eqs.(20) and (21) where  $Q(x)$  is the phase shift. On the basis of the above equations, it is found that the phase shift introduced by the shifter can be expressed by:

$$\theta = \frac{1}{ab} \frac{\epsilon - 1}{\sqrt{1 - \omega_c^2/c^2}} \frac{\omega}{c} \sin^2 \frac{\pi x}{a} \int_{-\infty}^{+\infty} \ln |R| dx \quad (27)$$

where  $a$  and  $b$  are the dimensions of the waveguide and  $x$  is the distance between the plate of the phase-shifter and the narrow wall of the guide. The reflection coefficient of the phase-shifter can be expressed by:

$$R = \frac{1}{ab} \sin^2 \frac{\pi x}{a} \frac{\omega}{c} \left[ \frac{\epsilon - 1}{\sqrt{1 - \omega_c^2/c^2}} \right] \int_{-\infty}^{+\infty} \ln |R| e^{-2ikz} dz \quad (28)$$

which, for a symmetrical plate, is in the form of Eq.(19). Eqs.(27) and (28) can be regarded as the basic formulae for

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Waveguide Phase-shifter Having a Low Reflection Coefficient

109-3-5-14/11

the form of Eq.(41). An experimental phase-shifter, described in Eq.(41), was constructed and it was found that its reflection coefficient was so low that it could not be measured by means of a measuring line. It was found by employing the method of Fig.5 that the standing wave ratio was better than 1.00. There are 5 figures and 12 references, 9 of which are Soviet and 3 English.

ASSOCIATION: Vsesoyuznyy n.-i institut fiziko-tekhnicheskikh i radioelektricheskikh izmereniy (All-Union Scientific Research Institute for Physico-engineering and Radio-engineering Measurements)

SUBMITTED: July 30, 1956

AVAILABLE: Library of Congress

Card 5/5

1. Wave ratio-Measurement 2. Phase shifter-Applications

AUTHOR: Gertsenshteyn, P. Ye.

1977/10-11-12/1

TITLE: Spatial Beats of Noise Waves in Coupled Delay Devices (Lines) (Prostranstvennyye lyeniya shumovikh voln v svyazannykh zamedlitel'yakh)

PERIODICAL: Radiotekhnika i Elektronika, 1977, Vol 22, No 10, pp 1254 - 1263 (USSR)

ABSTRACT: The investigation of complex problems of wave propagation in electron beams or in electron gas can be approximately treated as a problem of formal electrodynamics, since the fundamental equations contain a permittivity operator  $\hat{\epsilon}$  for the electron gas. The Maxwell equations are therefore written in the form of Eqs.(1) where  $\mathbf{j}$  and  $\rho$  are the currents and charges which excite the field. The permittivity is a function of frequency  $\omega$  and of the wave number  $\mathbf{k}$ , i.e:

$$\hat{\epsilon} = \hat{\epsilon}(\omega, \mathbf{k}) \quad (1)$$

where the sign on top denotes an operator. When solving the electrodynamic equations, a number of difficulties arise; firstly, the excited (electron) and the electromagnetic field have a large number of the degrees of

00:31/6

Special Beams of Noise Waves in Complex Delay Devices (1968)

freedom; secondly, the meaning of  $j$  and  $\rho$  is not clear and the sources of noise are not taken into account. It is therefore necessary to consider the following kinetic equation for the distribution function  $f$ :

$$\frac{\partial f}{\partial t} + v \frac{\partial f}{\partial r} + \frac{e}{m} \left\{ E + \left[ \frac{v}{c} H \right] \right\} \frac{\partial f}{\partial v} = 0 \quad (6)$$

The electromagnetic fields have to satisfy the equation system (4). The oscillatory component of the distribution function  $\phi$  is expressed by Eq.(5), where  $\hat{L}$  is a linear operator, as defined by Eq.(6). From the theory of linear, differential equations (Ref 8), it follows that the solution of Eq.(5) in a fixed region of space can be represented as Eq.(7), where  $\psi_{sv}$  corresponds to free oscillations and is independent of the field,  $\psi_{vyn}$  is analogous to the forced oscillations and proportional to the right-hand side of Eq.(5).

Card2/6

Consequently, the second component of the field  $\vec{E}_2$  is given by the boundary conditions given by Eq. (10), which describes the motion of electrons under the influence of the electric field components of the electromagnetic field. The solution of Eqs. (1) can be written as Eq. (9), where  $\vec{E}_2$  is the Green function of Eq. (5). The interaction components of the charges and currents can be expressed in Eq. (11), so that the component  $j_{\text{vyn}}$  can be linearly expressed by the field  $\vec{E}_2$ , as shown in Eq. (12). The operator  $\hat{G}$  is given by Eq. (13). The shot noise in electron beams can be evaluated on the basis of the distribution function given by Eq. (15), in which  $r_i$  and  $v_i$  are the radius vector and velocity of the  $i$ -th electron. On the other hand, the component of the distribution function expressed by Eq. (3), which describes the noise, is expressed in Eq. (16). Consequently, the function  $\psi$  can be expressed by analogy with Eq. (7), as a sum of two components (Eq. (17)). The current and the charge noise can be expressed in Eqs. (18), where  $\beta_1$  is the noise current and  $\beta_2$  is

0-103/6





Spatial Scats of noise Waves in Coupled Drive D Vlasov (111.0)

There are 17 references, 15 of which are Soviet, 1 English and 1 German; three of the Soviet references are translated from English.

ASSOCIATION: Vsesoyuznyy n.-i in-t fiziko-tekhnicheskikh izm.  
Radioelektronicheskikh izm. (All-Union Scientific  
Research Institute of Physico-technical and  
Radio-engineering Measurements)

SUBMITTED: July 30, 1956

Card 6/6

1. Delay lines--Theory      2. Electromagnetic fields--Mathematical analysis

6(4), 7(7)

SV. 108-13-12-3/12

AUTHORS: Gertsenshteyn, M. Ye., Pokrov, A. M., Solov'yev, L. G.

TITLE: Multi-Channel System of Parallel Selection Waveguides with  
Variable Couplings (Mnogostvol'naya sistema parallel'noy  
selektzii s reguliruyemyimi svyazjami)

PERIODICAL: Radiotekhnika, 1958, Vol 13, Nr 12, pp 20-25 (USSR)

ABSTRACT: With relatively narrow bands or not too high claims with respect to the adaptation, the problem of dividing or joining the channels can be solved by means of a system of shunted series-resonance circuits. The various filters are connected, in parallel to each other, to the common conductor by a simple or compact tap. A simple method of setting up a tap for the shunted series-resonance circuits is given. This method is based on the calculation data without intricate experimental work. At first, the paralleling of the resonance circuits is investigated. The obtained formulae (3) and (5) show that the tap must be tuned jointly with the filter connected to it, with one element. The input resistance of filters with several elements is then investigated and it is shown that the mutual influence of the various channels is determined essentially by the input resona-

Card 1/2



U7 10A-13-12-3/12

Multi-Channel System of Parallel Selection Waveguides with  
Variable Couplings

tors. Therefore, the input resonators of the filters with several elements must also be tuned with the taps. The connection of the filters to the common line is then investigated. The connection to the main waveguide is made variable by means of screws with a steplike cross section. By means of the method given in this article, a simple waveguide tap is worked out for a system with shunted series-resonance circuits with an input transient wave factor of  $\approx 0.95$  in the middle of the band. There are 7 figures, 1 table, and 3 Soviet references.

SUBMITTED: June 1, 1957

Card 2/2

GERTSENSHTEYN, M. YE.

56-1-55/56

AUTHORS: Bonch-Bruyevich, V. L. , Gertsenshteyn, M. Ye.

TITLE: On the Theory of the Magnetic Susceptibility of Metals (K teorii magnitnoy vospriimchivosti metallov)

PERIODICAL: Zhurnal Eksperimental'noy i Teoreticheskoy Fiziki, 1958, Vol. 34, Nr 1, pp. 261 - 261 (USSR)

ABSTRACT: The magnetic susceptibility of the electron gas was recently (references 1, 2, 3) calculated with the taking into account of the distant Coulomb correlation. In this connection, however, only the susceptibility caused by the Fermi branch of the spectrum of excitations was taken into account. But the authors want to call attention to the fact that the Bose quanta of plasma vibrations also furnish a certain contribution to the susceptibility. It is true that these excitations are neutral and do not furnish any contribution to the current, but their energy depends on the field strength of the magnetic field  $H$  and therefore the plasma-quanta are "carriers of magnetism". At the usual temperatures the real plasma-quanta are practically not excited in metal, but their zero energy also depends on  $H$ . This leads, as shown here, to a plasma-diamagnetism comparable with the Landau diamagnetism. In a weak magnetic field a separation of the plasma vibrations in longitu-

Card 1/2

56-1-55/56

On the Theory of the Magnetic Susceptibility of Metals

dinal and transversal vibrations is also possible. For the case discussed here only the former are of interest. An expression for the frequency of the longitudinal plasma-quantum is given. Then the author gives an expression for the magnetic susceptibility caused by the dependence of the zero energy of the plasma on the magnetic field. The neglect of the zero energy of the plasma is generally not at all justified and the quantitative agreement of the theory by Pines (reference 1) with the experiment must anew be checked. There are 5 references, 2 of which are Slavic.

ASSOCIATION: **Moscow State University**  
(Moskovskiy gosudarstvennyy universitet)

SUBMITTED: November 21, 1957

AVAILABLE: Library of Congress

Card 2/2

[illegible]

report submitted for the Centennial Meeting of the Scientific Technological Society of  
Radio Engineering and Electrical Communications by A. S. Popov (VSEIN), Moscow,  
6-12 June, 1959

AUTHOR: Gertsenshteyn, M.Ye.

SC7/10/94-1-27/50

TITLE Noise in an Electron Beam (O shumakh elektronnoye puchka)

PERIODICAL: Radiotekhnika i Elektronika, 1964, Vol 4, Nr 1,  
pp 146 - 147 (USSR)

ABSTRACT An electron beam contains two types of noise; one of these can be referred to as the cathode noise and is due to the emission processes at the cathode which produces the beam. The second type of noise can be referred to as the volume noise and is due to the processes occurring in the electron beam itself. In the vicinity of the cathode, the cathode noise is predominant while the volume noise is comparatively low. It can be expected that at large distances from the cathode, the volume noise will become significant, while the cathode noise is negligible. It is shown that the conditions for the predominance of the volume noise can be expressed by.

$$\gamma \sim 0.3 - 0.4 \quad (6)$$

$$\omega_0 \tau \gtrsim 0.2 - 0.5 \quad (7)$$

CS141/2

Noise in an Electron Beam

SOV/109-4-1-27/30

where  $\xi$  is given by Eq (5),  $\gamma$  is the transit time for the drift space and  $\omega_c$  is the Langmuir frequency.

In Eq (5),  $u_0$  is the electron beam velocity,  $v_0$  is the thermal electron velocity and  $\omega$  is the operating frequency.

There are 3 references, 2 of which are Soviet and 1 English.

SUBMITTED. February 21, 1958

Card 2/2

16(1),16(2)

AUTHORS: Gertsenshteyn, M. Ye., and Vasil'yev, V. B. 05/93  
307/52-4-4 4/13

TITLE: Waveguide With the Random Inhomogeneities and Brownian Motion  
on the Lobachevskiy Plane

PERIODICAL: Teoriya veroyatnostey i yeye primeneniya 1958  
Vol 4, Nr 4, pp 424-432 (USSR)

ABSTRACT: The authors consider a waveguide with random inhomogeneities. Let  $r_1$  be the reflection coefficient (ratio of the amplitudes of the reflected and original wave) of a single inhomogeneity. Let all  $r_1$  be independent random functions with known statistical characteristics. The authors ask for the reflection coefficient of the whole waveguide. It is shown that the problem can be reduced to the Brownian motion in the Lobachevskiy plane. At first two inhomogeneities are considered and it is stated that the resulting reflection coefficient is a bijective linear function mapping the unit circle onto itself. Therewith the relation with the Lobachevskiy plane is given. For several inhomogeneities the image point moves in the Lobachevskiy plane, while the sum of the corresponding noneuclidean distances yields the total effect of the inhomogeneities. If the considered random process is continuous, then it leads to the diffusion equation in the Lobachevskiy plane.

SUBMITTED: December 25, 1958  
Card 1/1

AUTHORS: Gertsenshteyn, M.Ye. and Vasil'yev, V.B. SOV/109-4-4-7/24  
 TITLE: The Diffusion Equation of a Statistically Non-homogeneous Waveguide (Diffuzionnoye uravneniye dlya statisticheski neodnorodnogo volnovoda)  
 PERIODICAL: Radiotekhnika i elektronika, 1959, Vol 4, Nr 4, pp 611 - 617 (USSR)  
 ABSTRACT: It is assumed that the complex reflection coefficient of the system is  $r = x + iy$  and that its probability density distribution satisfies the diffusion equation:

$$D \left( \frac{\partial^2 W}{\partial x^2} + \frac{\partial^2 W}{\partial y^2} \right) = \frac{\partial W}{\partial z} \quad (3)$$

where  $D$  is the statistical characteristic of the waveguide; this is equal to the average half sum of the reflection coefficients squared per unit length of the waveguide;  $z$  is the distance along the length of the waveguide. If a normalised variable  $t = \int D dz$  is introduced. the equation can be written as Eq (4). When the waveguide

Card1/4



SOV/10)-1-4-7/24

The Diffusion Equation of a Statistically Non-homogeneous Waveguide

is terminated with a matched load, the solution of Eq (4) is in the form of Eq (5). It is seen that for large  $t$ , Eq (5) has no physical meaning. A different differential equation for the density probability function is, therefore, necessary. The equation should be such as to make the solution independent of the terminating load: also when  $x^2 + y^2 \rightarrow 0$ , the differential equation should coincide with Eq (4). These requirements are satisfied by

$$\Delta W = - \frac{\partial W}{\partial t} \quad (7)$$

where  $\Delta$  is the Laplace operator on the Lobachevskiy surface. The operator is defined by Eq (8). By introducing a polar system of co-ordinates  $\eta, \phi$ , as defined by Eqs (9), the Laplace operator is represented by Eq (10). If  $\eta = i\theta$  and  $u = \text{ch}\eta$ , Eq (10) can be expressed as Eq (11). This can be solved by introducing the Laplace transformations and leads to the Legendre equation which

Card2/4

SOV/109-4-1-7/24

The Diffusion Equation of a Statistically Non-homogeneous Waveguide

is in the form of Eq (13). In its final form, Eq (13) can be written as Eq (16). On the basis of the above, it is found that the average value for  $u$  is expressed by Eq (17). The average value of the reflection coefficient is approximately expressed by Eq (19). The value of the average reflection coefficient  $r$  as a function of  $t$  is plotted in Figure 2; Curve I corresponds to a linear approximation, while Curve II represents more accurate results. It is seen that Curve I gives values which are higher than those represented by Curve II. The physical meaning of this is that a part of the energy of the reflected wave travelling from the load towards the generator is reflected by the non-uniformities of the waveguide (towards the terminating load). The authors make acknowledgment to B.Ye. Kinber for discussing the work and for his valuable remarks.

Card 5/4

SOV/102-4-4-7/24

The Diffusion Equation of a Statistically Non-homogeneous Waveguide

There are 2 figures and 9 references, 1 of which is English and 8 Soviet. 1 of the Soviet references is translated from English.

SUBMITTED: November 26, 1957

Card 4/4



Possible Use of Manned Balloons for  
Intelligence - Discovered and  
Controlled

1. Introduction

2. Background

3. Discussion  
4. Conclusion

5. References

6. Appendix

7. Notes

End

Possibility of Measuring the Velocity of  
Gravitational Distribution under Laboratory  
Conditions

1954  
SOV/55-37-5-54/11

gravity. There is a Soviet reference.

SUBMITTED: July 29, 1954

Card 3/3

GERTSENSTEYN, M. Ye.; BRYANSKIY, L.N.

Using phase shifters for eliminating mismatch errors. Izv.tekh.  
no.1:48-51 Ja '60. (MIRA 13:5)

(Phase converters)

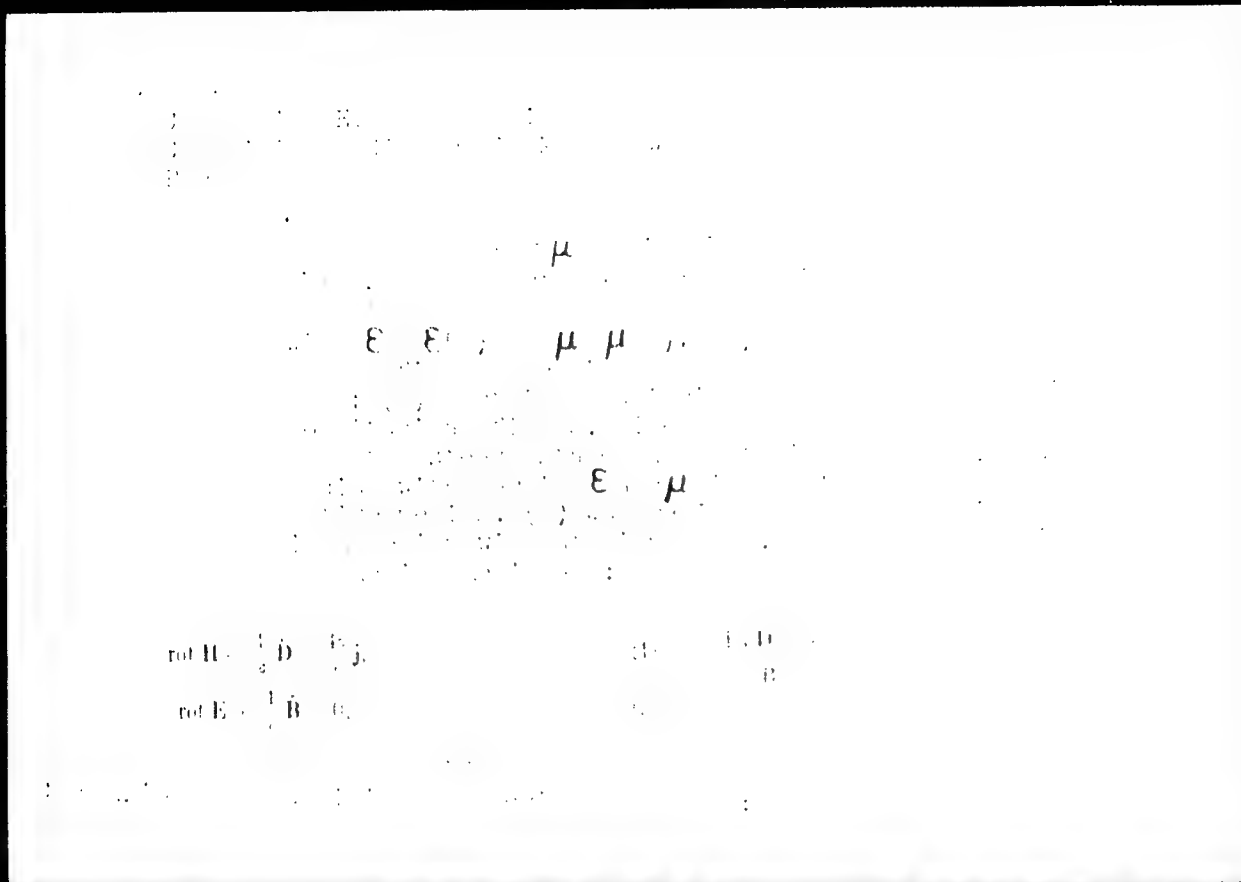
ATTORNS: Dr. J. Edgar Hoover, A. E. J. ...

NAME: J. Edgar Hoover, Director, Federal Bureau of Investigation

PERIOD: 1940-1941 (1941)

ABSTRACT: This is a copy of a letterhead memorandum (LHM) dated 10/10/40, from the Director of the Federal Bureau of Investigation (FBI) to the Chief of the New York Office, regarding the activities of the American Communist Party (CP) in New York City. The LHM is classified as "Confidential" and is marked with a large "X" in the top right corner. The text of the LHM discusses the CP's efforts to recruit and organize in New York City, and the FBI's response to these activities. The LHM is signed by J. Edgar Hoover, Director of the FBI.





Consider Electrodynamics in a Region  
 Containing a Dielectric Medium with  
 Variable Permittivity

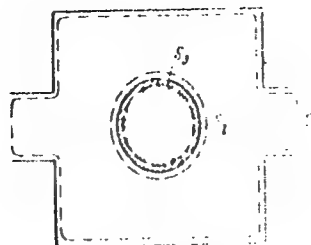
$$\mathbf{D} = \epsilon(t) \mathbf{E}, \quad (1)$$

$$\mathbf{B} = \mu(t) \mathbf{H}, \quad (2)$$

$\mathbf{E}$  and  $\mathbf{H}$  are weak solutions of the system of equations  
 describing the electromagnetic field in a region with  
 variable permittivity. The system of equations is written in  
 the form:

$$-\frac{1}{4\pi} \operatorname{div}(\mathbf{E}\mathbf{H}) = \frac{1}{8\pi} \frac{d}{dt} (\epsilon \mathbf{E}, \mathbf{E}) + \frac{1}{8\pi} \frac{d}{dt} (\mu \mathbf{H}, \mathbf{H}) = 0$$

where the terms in parentheses are the energy of the  
 electromagnetic field. The system of equations is written in  
 the form of a variational problem.



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 (5) ...  
 (8) ...  
 ... (S) ...  
 ... of the ...

$$P_1 = \frac{1}{S_1} \int_{S_1} (E \cdot H) \cdot dS$$

$$P_2 = \frac{1}{S_2} \int_{S_2} (E \cdot H) \cdot dS$$

Forced Oscillations of a Resonant System in a  
 Uniformly Anisotropic Medium with Time-Dependent  
 Parameters

It is shown that the forced oscillations of a resonant system in a uniformly anisotropic medium with time-dependent parameters are described by the following equation:

$$P_1 - P_2 = \frac{d}{dt} \frac{1}{8\pi} \int_V (\epsilon E, E) + (g H, H) dV + \frac{1}{8\pi} \int_V (\dot{\epsilon} E, E) + (\dot{g} H, H) dV = 0$$

The above is valid for a resonant system in a medium with time-dependent parameters. While the field vectors  $E$  and  $H$  are unknown, the relations (1) can be considered as an approximate solution of the problem. Relations (1) are obtained for a resonant system with self-excitation. The relations of normal waves in a medium with time-dependent parameters are also valid. The analysis of the forced oscillations of a resonant particle in a medium with time-dependent parameters where  $\epsilon = \epsilon(t)$ ,  $\mu = \mu(t)$ ,  $g = g(t)$  is also valid. The analysis of the forced oscillations of a resonant particle in a medium with time-dependent parameters where  $\epsilon = \epsilon(t)$ ,  $\mu = \mu(t)$ ,  $g = g(t)$  is also valid.

Let us consider a medium with a constant permittivity  $\epsilon_0$  and permeability  $\mu_0$ . The electric field  $E$  and magnetic field  $H$  are related by the following equations:

$$E = E_0 + E_1(t), \quad (1)$$

$$\mu = \mu_0 + \mu_1(t).$$

The electric field  $E$  and magnetic field  $H$  are related by the following equations:

$$E = \sum_i a_i(t) E_i(r), \quad (2)$$

$$H = \sum_i b_i(t) H_i(r). \quad (3)$$

A relation of order  $n$  between  $E$  and  $H$  is given by the following equation:  $E = \sum_i a_i(t) E_i(r)$ ,  $H = \sum_i b_i(t) H_i(r)$ . The relation  $E = \sum_i a_i(t) E_i(r)$  is given by the following equation:  $E = \sum_i a_i(t) E_i(r)$ .

$$\text{rot } E_q = \frac{1}{c} \omega_q E_q, \quad (4)$$

$$\text{rot } H_q = \frac{1}{c} \omega_q E_q. \quad (5)$$

and 6/11

the 1990s, the number of people in the world who are illiterate has increased from 1.2 billion to 1.5 billion. The number of illiterate people in the world is expected to reach 1.7 billion by the year 2015. The number of illiterate people in the world is expected to reach 1.7 billion by the year 2015.

$$\begin{aligned} i \sum_q \omega_q b_q z^n E_q &= \sum_q \left[ i z^n + \frac{d}{dt} \left( \frac{z^n}{\omega_q} \right) \right] E_q = -i z^n \sum_q \omega_q E_q \\ i \sum_q a_q \omega_q z^n H_q &= \sum_q \left[ i z^n + \frac{d}{dt} \left( \frac{z^n}{\omega_q} \right) \right] H_q = -i z^n \sum_q \omega_q H_q \end{aligned}$$

[illegible]

$$\dot{D}_1^{(m)} = \dot{d}_1 + \sum_{j=1}^m \int_0^1 \dot{\varphi}_j(\alpha) d\alpha,$$

$$w_{ij} = \frac{1}{n} \sum_{k=1}^n w_{ijk} \quad (1)$$

at

$$\begin{aligned} \dot{x}_p(t) &= \dot{U}_p(t) + \dot{F}_p(t) + \dot{G}_p(t) + \dot{H}_p(t) \\ \dot{x}_p(t) &= \dot{U}_p(t) + \dot{F}_p(t) + \dot{G}_p(t) + \dot{H}_p(t) \end{aligned}$$

where  $\dot{U}_p(t)$  is the derivative of the function  $U_p(t)$  with respect to time  $t$ ,  $\dot{F}_p(t)$  is the derivative of the function  $F_p(t)$  with respect to time  $t$ ,  $\dot{G}_p(t)$  is the derivative of the function  $G_p(t)$  with respect to time  $t$ , and  $\dot{H}_p(t)$  is the derivative of the function  $H_p(t)$  with respect to time  $t$ .

$$\begin{aligned} \dot{x}_p(t) &= \dot{U}_p(t) + \dot{F}_p(t) + \dot{G}_p(t) + \dot{H}_p(t) \\ \dot{x}_p(t) &= \dot{U}_p(t) + \dot{F}_p(t) + \dot{G}_p(t) + \dot{H}_p(t) \end{aligned} \quad (17)$$

Conformal Electrostatics of a R  
Containing A Dielectric Medium With  
Variable Parameters

In a similar way, the problem of the electrostatic field in a medium with variable parameters can be transformed into two equivalent problems, one in a medium with constant parameters  $\epsilon_0$  and  $\mu_0$  only. (1) Consider the problem of the electrostatic field in a medium with variable parameters  $\epsilon(\omega, t)$  and  $\mu(\omega, t)$ . The normal waves in a medium with variable parameters  $\epsilon(\omega, t)$  and  $\mu(\omega, t)$  are given by the following equations:

$$\epsilon_z(\omega, t) = \sum_{n=1}^{\infty} \epsilon_{zn}(\omega) e^{i\omega_n t},$$

for

$$\mu_z(\omega, t) = \sum_{n=1}^{\infty} \mu_{zn}(\omega) e^{i\omega_n t}.$$

By using the method of separation of variables, the problem of the electrostatic field in a medium with variable parameters  $\epsilon(\omega, t)$  and  $\mu(\omega, t)$  can be transformed into two equivalent problems, one in a medium with constant parameters  $\epsilon_0$  and  $\mu_0$  only, and the other in a medium with variable parameters  $\epsilon(\omega, t)$  and  $\mu(\omega, t)$  only.

The problem of the electrostatic field in a medium with constant parameters  $\epsilon_0$  and  $\mu_0$  only can be transformed into two equivalent problems, one in a medium with constant parameters  $\epsilon_0$  and  $\mu_0$  only, and the other in a medium with variable parameters  $\epsilon(\omega, t)$  and  $\mu(\omega, t)$  only.

The problem of the electrostatic field in a medium with variable parameters  $\epsilon(\omega, t)$  and  $\mu(\omega, t)$  only can be transformed into two equivalent problems, one in a medium with constant parameters  $\epsilon_0$  and  $\mu_0$  only, and the other in a medium with variable parameters  $\epsilon(\omega, t)$  and  $\mu(\omega, t)$  only.

The problem of the electrostatic field in a medium with constant parameters  $\epsilon_0$  and  $\mu_0$  only can be transformed into two equivalent problems, one in a medium with constant parameters  $\epsilon_0$  and  $\mu_0$  only, and the other in a medium with variable parameters  $\epsilon(\omega, t)$  and  $\mu(\omega, t)$  only.



...  $\Delta$  ...  $\omega$  ...  $\mathbf{E} \cdot \mathbf{D}$  ...  $\mathbf{B} \cdot \mathbf{H}$  ...

$$\mathbf{E} = \sum_n \mathbf{E}_n e^{i\omega_n t}$$

$$\mathbf{H} = \sum_n \mathbf{H}_n e^{i\omega_n t}$$

$$\overline{\text{div}[\mathbf{E}\mathbf{H}]} = \frac{1}{V} \sum_n \sum_m \omega_n \omega_m (\mathbf{E}_n \cdot \mathbf{E}_m + \mathbf{H}_n \cdot \mathbf{H}_m)$$

...  $\mathbf{E} \cdot \mathbf{D}$  ...  $\mathbf{B} \cdot \mathbf{H}$  ...

...  $\mathbf{E} \cdot \mathbf{D}$  ...



Let  $\mathcal{H}$  be a Hilbert space and  $\mathcal{A}$  a self-adjoint algebra of bounded operators on  $\mathcal{H}$ . Let  $\mathcal{P}$  be a projection in  $\mathcal{A}$ .

Let  $\mathcal{H}_1$  and  $\mathcal{H}_2$  be Hilbert spaces and  $\mathcal{A}_1$  and  $\mathcal{A}_2$  self-adjoint algebras of bounded operators on  $\mathcal{H}_1$  and  $\mathcal{H}_2$  respectively. Let  $\mathcal{P}_1$  and  $\mathcal{P}_2$  be projections in  $\mathcal{A}_1$  and  $\mathcal{A}_2$  respectively.

$$\Psi: \quad \Pi = -\operatorname{grad} \mathcal{U}$$

Let  $\mathcal{H}$  be a Hilbert space and  $\mathcal{A}$  a self-adjoint algebra of bounded operators on  $\mathcal{H}$ .

$$-\operatorname{div} p \operatorname{grad} \mathcal{U} = 2 \left( \frac{\partial \mathcal{U}}{\partial x_1} \frac{\partial \mathcal{U}}{\partial x_2} + \frac{\partial \mathcal{U}}{\partial x_3} \frac{\partial \mathcal{U}}{\partial x_4} \right)$$

Let  $\mathcal{H}$  be a Hilbert space and  $\mathcal{A}$  a self-adjoint algebra of bounded operators on  $\mathcal{H}$ . Let  $\mathcal{P}$  be a projection in  $\mathcal{A}$ .

$$\frac{\partial^2 \mathcal{U}}{\partial x_1^2} + (1+k) \left( \frac{\partial^2 \mathcal{U}}{\partial x_1 \partial x_2} + \frac{\partial^2 \mathcal{U}}{\partial x_2 \partial x_1} \right) = 0$$

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$$\Delta T = \frac{\partial T}{\partial x} + \frac{\partial T}{\partial y} + \frac{\partial T}{\partial z} = 0$$

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$$S = \frac{c}{4\pi} [EH^*] - \frac{u}{8\pi} \frac{d_1}{dk} HH^*, \quad (34)$$

... ..  $k$  ... ..

$\mu_L$   $k$

... ..  $E$  ... ..



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SOV/109-5-2-5/26

AUTHOR: Gertsenshteyn, M. E.

TITLE: Phase and Frequency Distortions in Mixers

PERIODICAL: Radiotekhnika i elektronika, 1960, Vol 5, Nr 2,  
pp 214-217 (USSR)

ABSTRACT: Amplitude and phase distortions in crystal mixers at super high frequencies are analyzed assuming that the mixer is a six-pole network which can be described by a corresponding matrix of conductivity. This leads, however, to cumbersome calculations and not comprehensive end results. Provided the non-uniformity of the frequency characteristic is relatively mild, approximation methods can be used. The proposed method takes the wave picture as a starting point rather than currents and voltages. Distortions can be described by the interference of several waves arriving by different ways into the

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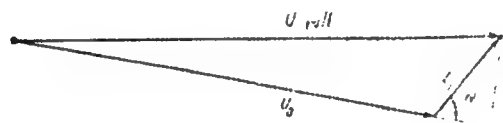
Phase and Frequency Distortions in Mixers

7/772

SCN/109-5-2-5/26

output field of the system. In the ideal s-h-f system only one path exists, but in the real system there may be parasitic paths caused by (a) detuning in the wave guide, (b) double conversion at mirror-frequency or due to harmonics, (c) "squeezing" of the signal from the oscillator to the receiver due to poor shielding. In all these cases an analysis of the frequency characteristics can be made by a vector analysis of the diagram. The full field at the output of the system is a vector sum (see Fig. 1.)

$$\vec{U}_{full} = \vec{U}_0 + \vec{U}_1 = U_0 \left( 1 + \frac{\vec{U}_1}{U_0} \right) \quad (1)$$



Card 2/9

Fig. 1.

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[illegible]

1111

$\gamma$

$$\Delta A(x) = 8,647 \cdot 10^{-9} ;$$

$$\Delta_T = 7 \cdot 10^{-9}.$$

[illegible]

1. *Phragmites* (common)



Figure 1. The effect of the concentration of the

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of the concentration of the (a)

of the concentration of the (b)

of the concentration of the (c)

of the concentration of the (d)

$$\begin{aligned} \left[ \Delta t \right] &= \frac{1}{\omega} \left[ \Delta \omega \right] \\ \left[ \Delta t \right] &= \frac{1}{\omega} \left[ \Delta \omega \right] \left( \frac{1}{\omega} \right) \\ \left[ \Delta t \right] &= \frac{1}{\omega} \left[ \Delta \omega \right] \left( \frac{1}{\omega} \right) \end{aligned} \quad (7)$$

Carl ...  
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 (will ...)

From the Department of Mathematics

University of California, San Diego

1970

$$\pi = \frac{d^2 \Delta z}{\omega^2} \Delta \omega \approx \frac{1}{\gamma} \left( \frac{d}{\omega} \right) \sin^2 \theta \approx \frac{d^2}{\omega^2} \sin^2 \theta$$

where  $\gamma = \frac{c}{v}$

$$\omega = \frac{2\pi}{\lambda} \sin \theta \left( \frac{d}{\omega} \right) \approx \frac{2\pi}{\lambda} \left( \frac{d}{\omega} \right) \sin \theta \approx \frac{2\pi}{\lambda} \left( \frac{d}{\omega} \right) \sin \theta$$

Applying the above results to the case of a plane wave incident on a grating, we find that the diffraction angle  $\theta$  is given by

$$\sin \theta = \frac{\lambda}{d} \left( \frac{d}{\omega} \right) \approx \frac{\lambda}{d} \left( \frac{d}{\omega} \right) \sin \theta$$

Consequently, the diffraction angle  $\theta$  is given by  $\sin \theta = \frac{\lambda}{d} \left( \frac{d}{\omega} \right) \approx \frac{\lambda}{d} \left( \frac{d}{\omega} \right) \sin \theta$

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100

1. The first part of the paper is devoted to the study of the asymptotic behavior of the solutions of the system (1) as  $\epsilon \rightarrow 0$ . It is shown that the solutions of the system (1) converge to the solutions of the system (2) in the sense of the weak convergence in the space  $L^2(\Omega; \mathbb{R}^n)$ .

**Figure 1**

(c) Derivatives of  $\mathbf{g}$  and  $\mathbf{h}$  are  $\mathbf{g}'(t) = \mathbf{g}(t) \mathbf{g}^T(t)$  and  $\mathbf{h}'(t) = \mathbf{h}(t) \mathbf{h}^T(t)$ . If  $\mathbf{g}$  and  $\mathbf{h}$  are both positive semidefinite, then the product  $\mathbf{g}'(t) \mathbf{h}(t)$  is also positive semidefinite. The product  $\mathbf{h}'(t) \mathbf{g}(t)$  is also positive semidefinite and also symmetric. By (a), (b) can be written as

$$(6.54) \quad \sum_{i=1}^n \frac{1}{\lambda_i} \left( \frac{\partial}{\partial \lambda_i} \log \lambda_i \right) = 0 \quad \text{for } \lambda_i \neq 0.$$

where, fourth, we note that the  $\delta$  is a small number,  $\delta \ll 1$ , and the  $\delta$  is a small number,  $\delta \ll 1$ .

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0.1 nsec,  $\Delta\omega/\omega \approx 10^{-4}$ ,  $\Delta\tau_{\text{eff}} \approx 10^{-10}$  sec.

Figure Frequency Distribution in M' and

$$\frac{1}{\sqrt{2\pi}} \int_{-\infty}^{\infty} f(x) e^{-j\omega x} dx$$

where, in addition,  $f(x)$  is the frequency distribution; the term in ( ),  $f(x)$  is the frequency distribution.

$$f(x) = \frac{1}{\sqrt{2\pi}} \int_{-\infty}^{\infty} f(x) e^{-j\omega x} dx \quad (1)$$

It is known that the frequency distribution  $f(x)$  is the Fourier transform of the power spectrum  $P(\omega)$  and the power spectrum  $P(\omega)$  is the Fourier transform of the frequency distribution  $f(x)$ .

$$P(\omega) = \int_{-\infty}^{\infty} f(x) e^{-j\omega x} dx \quad (2)$$

It is also known that the power spectrum  $P(\omega)$  is the Fourier transform of the frequency distribution  $f(x)$  and the frequency distribution  $f(x)$  is the Fourier transform of the power spectrum  $P(\omega)$ .

$$f(x) = \frac{1}{\sqrt{2\pi}} \int_{-\infty}^{\infty} P(\omega) e^{j\omega x} d\omega \quad (3)$$

Phase Frequency Distortion in Mixers

1971

Thus distortions due to phase shift in intermediate frequency circuits can be minimized or eliminated when due to incident phase shift. To avoid distortion at the output of the mixer, the frequency spectrum must be flat with respect to phase with a minimum of possible distortion. The phase of the signal must be placed after a traveling wave device, which is of no importance, the use of traveling wave devices is recommended. In the author's view, the author reiterates the recommendation of traveling wave devices between the mixer and the preceding stage of active resistors in order to prevent distortion. There are 4 B. N. references.

SUBMITTED: February 19, 1971

Card 1/1

2.3246

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307/1 1-3-7- 1-1

**AUTHOR S:** Gertsenshteyn, M. E., Kliner, N. E.

TITLE: Re: Subjectivity of Non-Religious Persons in Anti-  
Religion

PERIODICAL: Radiot. Elektron. i Elektronika, Moscow, Vol. 1, No. 1, 1968, pp. 1-10 (USSR)

**ABSTRACT:** In contrast to non-saturable oscillators, in phase-locked oscillators a saturated oscillation is maintained in a resonant system by feedback during the action of a periodic excitation. Therefore, one can say that a periodic amplifier will operate on the principle of phase-locked amplification. Let it be called a phase-locked amplifier. The need to deal with phase-locked amplifiers as a particular amplifier with one degree of freedom (with reference to amplified signal). The point is that there is a certain kind of distortion, which does not depend on the level of the excitation, which is not possible in the case of the operation of a non-saturable amplifier. The results of the calculations show that the phase-locked amplifier is characterized by a certain kind of distortion, which does not depend on the level of the excitation, which is not possible in the case of the operation of a non-saturable amplifier.

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$$\ddot{u} + 2\dot{u}\dot{\theta} + \omega^2(1 + q \sin \theta) u = \omega_0^2 \cos \theta t, \quad (1)$$

where  $\theta(t)$  is the angle of rotation of the pendulum about the vertical axis  $O$ . The angle  $\theta$  is measured from the vertical axis to the position of the pendulum bob. The frequency of the external force is  $\omega_0$ . The frequency of the pendulum is  $\omega$ . The parameter  $q$  is the ratio of the length of the pendulum to the radius of the Earth. The parameter  $\omega_0$  is the frequency of the external force. The parameter  $\omega$  is the frequency of the pendulum.

... ..



Phonetic transcription of the  
Russian text is given in the

Appendix

1. The first part of the text is a list of the names of the  
members of the committee, which is headed by the  
Chairman, the Vice-Chairman, and the members.  
The names are listed in the following order: the  
Chairman, the Vice-Chairman, and the members.

2. The second part of the text is a list of the

members of the committee, which is headed by the  
Chairman, the Vice-Chairman, and the members.  
The names are listed in the following order: the  
Chairman, the Vice-Chairman, and the members.

3. The third part of the text is a list of the

1. The function  $f(x)$  is defined by the equation

$$f(x) = \frac{1}{2} \left( \varphi(x) + \frac{1}{\varphi(x)} \right) \quad (1)$$

where  $\varphi(x)$  is a function satisfying the condition

$$\varphi(x) = \frac{\pi}{\nu - \Omega} \quad (2)$$

0.14/1.



1. The first part of the paper is devoted to the study of the properties of the function  $\chi(\tau)$  defined by the equation

$$\chi(\tau) = \chi(\tau) + \dots + \chi(\tau) \quad (1)$$

$$\chi(\tau) = \chi(\tau) + \dots + \chi(\tau) \quad (2)$$

$$\begin{aligned} & \chi(\tau) = \chi(\tau) + \dots + \chi(\tau) \\ & \chi(\tau) = \chi(\tau) + \dots + \chi(\tau) \end{aligned} \quad (3)$$

$$\chi(\tau) = \chi(\tau) + \dots + \chi(\tau) \quad (4)$$

2. The second part of the paper is devoted to the study of the properties of the function  $\chi(\tau)$  defined by the equation

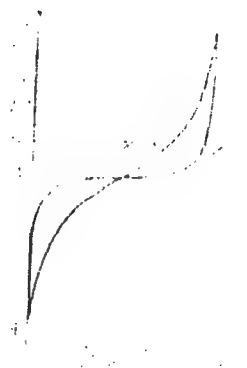
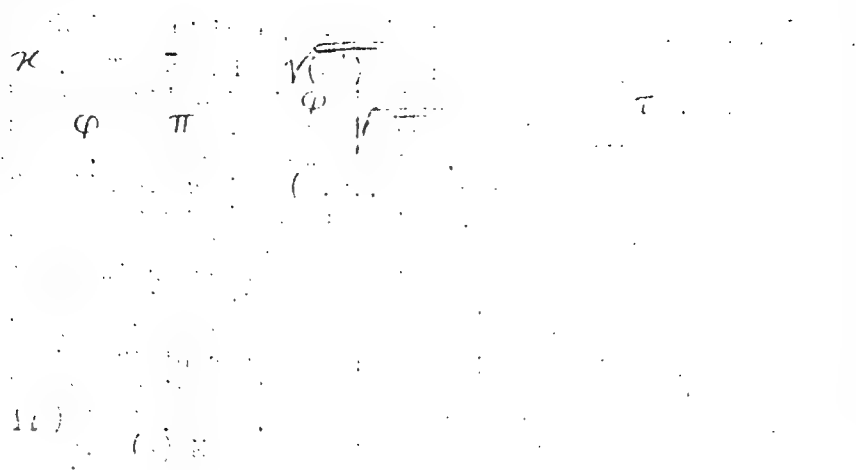


Figure 1. Theoretical curves of the  
theoretical curves of the



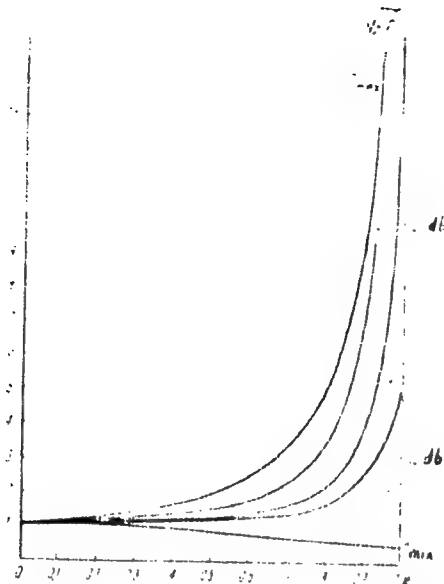


Fig. 5

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Please Call it in, or Single-Circuit  
 Program for Amplifier

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2001-10-10

and the term with negative frequency,  $\omega \sim -\omega$ , is  
 excluded. Assuming that only one side frequency  
 are essential: positive  $\Omega \approx \omega$  and negative  $\Omega - \nu$   
 $-\mu$ , for  $\nu \sim \omega$ ;  $|\mu| \sim \omega$ . The  
 fields are sought in combination. The  
 the product  $\Omega$  and  $\Omega - \nu$ :

$$y = a e^{i(\Omega - \nu)t} + b e^{i(\Omega - \nu)t}$$

where

$$a = \nu / (\Omega - \nu) \quad b = \mu / (\Omega - \nu)$$

a and b are the constants which are  
 specified in shown in Fig. 1. The  
 and  $\Omega$  are constant,  $\mu$  and  $\nu$  are  
 one-half the product of  $\mu$  and  $\nu$ . The  
 sought in the form:

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Fig. 6

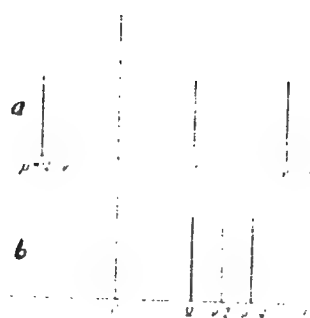


Fig. 6

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For the purpose of this report, the following  
assumptions were made:

1. The system is a linear, time-invariant system.  
2. The input signal is a unit impulse.  
3. The output signal is the system response.  
4. The system is described by the following differential equation:

$$R(s) = \frac{1}{s^2 + 2s + 1}$$

The transfer function of the system is given by:

$$H(s) = \frac{1}{s^2 + 2s + 1}$$

$$h(t) = \frac{1}{2} e^{-t} (1 - e^{-t})$$

End of Report





Phase Selectivity of Single-Circuit  
Parametric Amplifier

UDC 621.372.6.01  
621.372.6.01.02

$$I_1(t) = \int_{-\infty}^{+\infty} e^{i\omega t} \psi(\omega) d\omega$$

Thus, in the case of a signal of any shape, phase selectivity is also present. (6) Noise Amplification. White spectrum noise is the totality of incoherent sinusoids with arbitrary phases; their amplification coefficient is (11). With a quadratic indicator, it is:

$$|K^2| = \frac{1+x^2}{(1-x^2)^2} \quad (20)$$

Consequently, phase selectivity does not play any role in noise amplification. Equation (20) is also valid if not only the phase of the amplified signal, but also the phase of the pumping field is arbitrary (or at

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Phase Selectivity of Single-Circuit  
Parametric Amplifier

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random). It seems that the field of an incoherent source can be used as the pumping field. Noises and distortions of such an amplifier, of course, should be investigated separately. (7) Influence of Phase Selectivity. From the above, it follows that phase selectivity leads to amplitude and phase modulation of the signal being amplified. Pulses at the output of a parametric amplifier are amplitude-modulated. This modulation can be removed with the help of a system of automatic amplitude regulation in the receiver. Analyzing FM of the signal, the spectral method is recommended. Conclusions: (1) A parametric amplifier with one degree of freedom, when amplifying a signal with frequency  $\Omega$ , causes a beat modulation of the amplified signal, resulting in phase oscillations  $\nu = 2\Omega$ . (2) Solutions for near-resonance area by simplified equations and complex amplitude methods are identical, and the method of complex amplitudes can be used for the solution of more complicated problems. (3) A parametric amplifier with one degree of freedom is phase-selective, as its instant

Card 16/17



GERTSENSHTEYN, M. Ye.; VASIL'YEV, V.B.

In regards to S. I. Al'ber and V. I. Bespalov's letter "Diffusion  
equation for a statistically nonhomogenous wave guide. Radio-  
tekh. i elektron. 6 no.3:449-450 Mr '61. (MIRA 14:3)

(Wave guides)

(Al'ber, S. I.)

(Bespalov, V. I.)



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S/056/61/040/CO1/012/037  
B1C2/B204

24.4400

AUTHOR: Gertsenshteyn, M. Ye.

TITLE: The laws of conservation in the general relativity

PERIODICAL: Zhurnal eksperimental'noy i teoreticheskoy fiziki, v. 40,  
no. 1, 1961, 114-122

TEXT: The author deals with two points in the theory of the laws of conservation which are, seen from the mathematical viewpoint not clear: 1) The energy momentum vector  $P_i = \int t_i^k dS_k$  is in this integral representation not satisfactory, because the vector addition is not defined. 2) In the representation of the coordinate transformation (2):

$\delta x^i = \xi^i(x) = x_j^i(x) \delta \omega^j$ , where  $\delta \omega^j$  are the parameters of an element of the irreducible group of coordinate transformation (translation or rotation), it is not definitely said what functions  $\xi^i(x)$  correspond to the translation. Integrals like the one abovementioned occur in the general relativity when the laws of conservation are being studied. If  $t_i^k$  is an energy momen-

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tum pseudotensor, and if composition (integration) is carried out according to components (the coordinates are Euclidean at infinity), then the integral is independent of coordinate system; if, at  $x^\alpha \rightarrow \infty$ ,  $\xi^i$  tend  $\rightarrow$  const, the integral quantities, which were obtained in the integration of various energy-momentum tensors (which are produced by (2)), coincide. Such a situation, where the mathematical operation employed is not defined, and obtains sense only by the nature of the expression under the integral, is considered to be unsatisfactory by the author. Definition of the integral and the translation is purely geometric, and ought to be independent of the physical content of the problem. For determination of this integral in Riemann geometry, a so-called "free" vector field is introduced, which uniquely (i.e., independent of path) describing the shift of the origin of the coordinates is introduced:  $P_i(x) = \hat{C}P_i(x_0)$ , where  $\hat{C}$  is the operator of the "harmonic" shift.  $\xi^i(x)$  is considered to be a vector field, which obeys the following conditions:  $\xi^i(x, x_0)$  is a unique function,  $\xi^i(x)$  are vectors which are parallel in Euclidean space. Thus it is possible, like above, to put  $\xi^i(x) = \hat{C}\xi^i(x_0)$ . The harmonic shift is defined in all spaces

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that are topologically equivalent to Euclidean space; however, it differs from a parallel shift. For the "free" vector in a curvilinear pseudoeuclidean space  $\nabla_k P^s(x) = 0$  holds,  $\nabla_k$  denotes a covariant derivative, if  $k$  and  $s$  are independent, this equation contains 16 conditions. With the definition of the invariant  $\xi = \nabla_k P^k = \text{div} P$ , and separation of the symmetric and anti-symmetric part,  $\xi_{ik} = \nabla_k P_i + \nabla_i P_k$ ,  $\eta_{ik} = \nabla_k P_i - \nabla_i P_k$ , it is possible to impose onto the vector field  $P_i(x)$  the condition  $\xi_{ik} = 0$  (which in itself comprises 10 conditions). These conditions have already been studied by V. A. Fok. They are satisfied only in a space of constant curvature ( $\nabla_s R = 0$ ). The conditions (13):  $\xi = 0$ ,  $\eta_{ik} = 0$  (7 conditions) are, on the other hand, satisfied in the case of arbitrary  $R_{iks}^m$ . The solution of (13) is given with  $P_k = \nabla_k \varphi = \partial \varphi / \partial x^k$ ,  $\square \varphi = 0$ . The general-covariant linear differential equations (13) define the geometric operation of a "harmonic" translation

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of the vector in a unique manner. There now exists, also in the general case of a space of arbitrary curvature, a preferred system of coordinates, in which the components of the vector remain unchanged in the case of a shift. The condition

$$\partial(\sqrt{-g} g^{ik})/\partial x^i = 0; g^{im} \Gamma_{im}^k = 0 \text{ determines the class of the}$$

"harmonic" (preferred) system of coordinates. In such a system, the covariant vector components in harmonic translation do not change, and it is therefore possible to integrate the vectors by the components. Energy-momentum vector, - pseudotensor, energy density, and the Hamiltonian of the system should, therefore, be calculated in such a harmonic system. The case of infinitely small coordinate transformations is studied and the formula hereby for the energy-momentum tensor is applied to the gravitational field. For the canonic energy-momentum tensor, a unique expression is obtained which after symmetrization goes over into the Landau-Lifshits tensor. In conclusion, the case is studied in which the gravitational field may be considered to be a slight perturbation, and the results of calculations are compared in the various systems of coordinates. The

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author finally thanks V. L. Bonch-Bruyevich, Professor A. Z. Petrov,  
A. A. Fedorov, and L. G. Solovey for discussions. There are 10 references:  
4 Soviet-bloc and 4 non-Soviet-bloc.

SUBMITTED: October 8, 1959 (initially) and March 9, 1960 (after revision)

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26412  
S/056/64/341/001/007/021  
B102/B214

AUTHOR: Gertsenshteyn, M. Ye.

TITLE: Wave resonance of light and gravitational waves

PERIODICAL: Zhurnal eksperimental'noy i teoreticheskoy fiziki, v. 41,  
no. 1(7), 1961, 113-114

TEXT: This paper gives an estimate of the energy of gravitational waves produced during the propagation of light in a constant electric or magnetic field. According to general relativity light and gravitational waves propagate with equal velocity, and the corresponding rays coincide with the zero geodesics. That means that, if there exists a linear relationship between light and gravitation waves, wave resonance known in radio physics must appear so that even in weak coupling a significant energy transfer may take place. In the presence of an electromagnetic field a weak gravitational field is described by

$$\square \psi^A = -16\pi \gamma c^{-4} T^A, \quad \tau_A^A = 0, \quad \tau_{,A}^A = 0, \quad (1)$$

$$\tau_A^A = \frac{1}{4\pi} (F^{\mu\nu} F_{\mu\nu} - \frac{1}{2} \delta_A^A (F^{\mu\nu} F_{\mu\nu})), \quad \psi_A^A = h_A^A - \frac{1}{2} h \delta_A^A.$$

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where  $\tau^{ik}$  is the energy - momentum tensor of the electromagnetic field,  $F^{ik}$  is the electromagnetic field tensor,  $\gamma$  the gravitational constant, and  $h_{ik}$  the perturbation of the metric tensor. Eq. (1) is used for investigating the propagation of light ( $F^{ik}$  field) in the presence of a strong magnetizing field  $F^{(0)ik}$  constant in space and time. The energy - momentum tensor becomes the sum of three terms: square of a constant term, square of the light wave field, and an interference term describing the wave resonance. On neglecting the non-resonance term one obtains the relation

$$\square \Psi_k = - \frac{8\gamma}{c^2} \left[ F^{(0)ik} F_{ik} - \frac{1}{4} \delta_k^i (F^{(0)lm} F_{lm}) \right]. \quad (2).$$

If the  $y$ -axis is taken in the direction of the wave vector and the wave amplitude is expressed in the units of energy density, one obtains

$$F_{ki} = b(x) f_{ki} e^{ikx}, \quad f_{0i} f_{0i} = 1, \quad k = \omega/c, \quad (3)$$

$$\Psi^{ik} = a(x) \sqrt{16\pi\gamma/c^4 k^2} \zeta^{ik} e^{ikx}, \quad \zeta_{ik} \zeta^{ik} = 1, \quad \zeta_i^i = 0;$$

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where the amplitudes  $f_{kl}$  and  $i k$  are dimensionless. With this one obtains in the approximation of slowly varying amplitudes:  $ida(x)/dx$

$= \sqrt{\gamma/\pi c^4} F^{(0)il} f_{kl} \int_i^k b(x)$ . The solution of this equation has the form  $a(x) = i \sqrt{\gamma/\pi c^4} f_{kl} \int_i^k F^{(0)il}(s) \cdot b(s) ds + a(0)$ , where the integration

is made along the ray. If  $a(0) = 0$  the external field is constant and the absorption or scattering of the light along the ray is small in the domain considered; i.e.  $b(s) = \text{constant}$  so that  $|a(x)/b(0)|^2 = (\gamma/\pi c^2) F^{(0)2} T^2$ , where  $T$  is the time in which the ray traverses the constant field. The amplitude packet was here set equal to one. If the  $F^{(0)}$  field is turbulent and random, it can be assumed for the purpose of estimating the energy of

the gravitational wave that  $F^{(0)}$  is constant along a path of length  $R_0$  ( $R_0$  - correlation radius of the  $F^{(0)}$  field) and then changes by jumps and at random. The light amplitude  $b(x)$  is practically constant along the ray; the amplitude of the gravitational wave is given by

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$$a(x) = \sum a_n; \quad a_n = i \sqrt{\gamma/\pi c^4} f_n \zeta_n \int_{x_{n-1}}^{x_n} F^{(0)H}(s) b(s) ds.$$

The gravitational waves excited at each portion of the path become incoherent. One obtains:  $|a(x)/b|^2 = (\gamma/\pi c^3) F^{(0)2} R_0 T$ . (7). For

interstellar fields one obtains, for example,  $|a/b|^2 \sim 10^{-17}$ ,  $(T^{(0)} = 10^{-5} G$ ,  $R_0 = 10$  light years,  $T = 10^7$  years). The frequency of the excited

gravitational wave is determined by the light frequency. Strong magnetic fields exist also inside the stars, and therefore gravitational waves can be produced. Here the correlation radius  $a(x)$  is essentially determined by the free path of the radiation. For the calculation of the intensity of this wave (7) can also be used, but then  $T$  is the diffusion time of the energy of the ray in the star transparent to the radiation. It can be shown that (7) represents the ratio of the gravitational and light radiations of the star. Naturally, the intensity of the gravitational radiation is small and is unimportant for the energy balance of the star. There are 3 Soviet-bloc references.

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9.3240 (1040, 1139, 1154)

AUTHOR: Rabinovich-Vizel', A.A., and Gertsenshteyn, M.Ye.

TITLE: On the bandwidth of frequency multipliers employing non-linear capacitance

PERIODICAL: Radiotekhnika i elektronika, v. 7, no. 3, 1962.  
175 - 177

TEXT: The purpose of the paper is to determine the bandwidth of frequency multipliers using non-linear elements. The authors first survey available literature and conclude that the efficiency of this type of frequency multiplier has received much attention, but hardly anything has been written on the attainable bandwidth. Next they quote K.M. Johnson's formulas, slightly rearrange them and find for the product of relative bandwidth and optimum efficiency

$$\eta_{\text{opt}} \frac{\Delta f}{f} = \sqrt{b_n^2 + (\omega_1 \tau)^2} \cdot \omega_1 \tau. \quad (6)$$

where  $b_n$  depends on the nonlinear characteristics of the diode em-  
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On the bandwidth of frequency ...

ployed.  $\omega_1$  - fundamental frequency,  $\tau$  - time constant of the diode,  
n - factor of multiplication. For a lossless diode

$$\tau = 0, \eta = 1, \frac{\Delta f}{f} = b_n, (Q_{D1} b_n)^2 \gg 1 \quad (7)$$

where  $Q_{D1}$  - quality of the diode at the frequency  $\omega_1$ . In this case  
the bandwidth is dependent on n. If the losses are large  $Q_{D1} b_n \ll 1$ ,  
the bandwidth is mainly determined by the losses and independent of  
the harmonic number. If non-linear resistances are used there is no  
difficulty with bandwidth because broadband matching is possible.  
There are 5 references: 1 Soviet-bloc and 4 non-Soviet-bloc. The 4  
most recent references to the English-language publications read as  
follows: C.H. Page, J. Res. Nat. Bur. Standards, 1956, 56, 4, 179.  
G. Luetgenau, M.V. Duffin, and P.H. Dirnbach, IRE Wescon Convention  
Record, 1960, part 3, 13; P.M. Fitzgerald, T.H. Lee, M.S. Moy, E.J.  
Powers and J.J. Younger, IRE Wescon Convention Record, 1960, part 2  
43; K.M. Johnson, IRE Trans., 1960, MTT-8, 5, 625.

SUBMITTED: July 20, 1961  
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2/106/02/007/003/005/029  
0234/1702

AUTHORS: Gertsenshteyn, N.Ye., and Kimber, B.Ye  
TITLE: Stability of the super-regenerative regime of an amplifier with complex networks  
PERIODICAL: Radiotekhnika i elektronika, v. 7, no. 3, 1962, 397 - 403

TEXT: The authors formulate equations for a parametric amplifier with variable capacity without frequency transformation, considering it as an n-terminal network. For the case of a two-circuit non-degenerate regenerative amplifier, an equation of Hill's type is deduced from the general equations; the stability of the solutions is determined by that of the solutions of the corresponding homogeneous equation. It is found that if a complicated input filter is used, whose band is not much wider than that of the amplifier, the domains of stability depend essentially on the parameters of superization. The case of an input filter consisting of two equal links is considered as an example; the homogeneous equation is reduced

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